

PSET 5

$$1.) G(s) = \frac{5}{s(1.4s+1)(\frac{s}{3}+1)}$$

Keep DC gain constant

$$-140 = -90 - \tan^{-1}\left(\frac{\omega}{1.4}\right) - \tan^{-1}\left(\frac{\omega}{3}\right) + \sin^{-1}\left(\frac{\alpha-1}{\alpha+1}\right)$$

α has max value of 10 to keep noise rejection reasonable

$$-140 = -90 - \tan^{-1}\left(\frac{\omega}{1.4}\right) - \tan^{-1}\left(\frac{\omega}{3}\right) + 53^\circ$$

$\omega_c = 2.7$ max crossover

Need geometric mean of lead to be equal to ω_c .

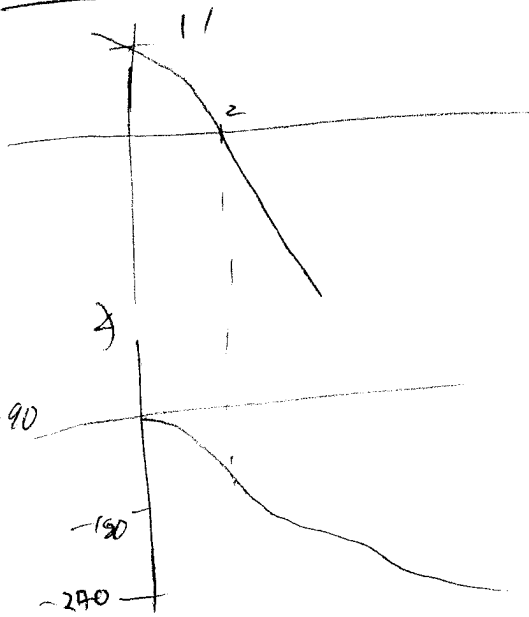
$$\omega_c = \frac{1}{\alpha \tau} \Rightarrow \tau = \frac{1}{\sqrt{10} (2.7)} \quad \tau = 0.1164$$

$$1 = K \frac{\left(\frac{10 \cdot 0.1164 \cdot 2.7}{3} + 1\right)^{1/2}}{\left(0.1164 \cdot 2.7 + 1\right)^{1/2}} \cdot \frac{5}{2.7 \left(\frac{2.7}{1.4} + 1\right)^{1/2} \left(\frac{2.7}{3} + 1\right)^{1/2}}$$

3.14633 1.633613

$$K = 1/2$$

$$\text{Lead} = \frac{1}{2} \left(\frac{10 \cdot 0.1164s + 1}{0.1164s + 1} \right)$$



$$\text{Lead} = \left(\frac{\alpha \tau s + 1}{\tau s + 1} \right)$$

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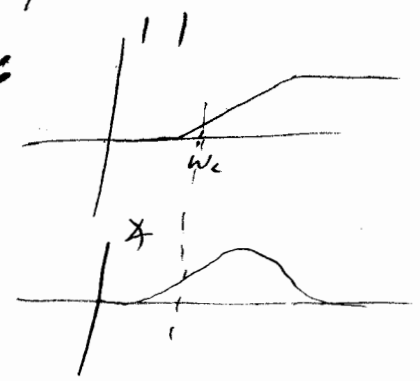
$$1.) \quad G(s) = \frac{5}{s(s/10 + 1)(s/3 + 1)}$$

$$\text{lead} = \frac{s/a + 1}{s/b + 1} \quad \text{with } a = 100 \text{ gives } 78^\circ \text{ of phase}$$

Placing zero at current crossover will keep crossover somewhat the same. However, the additional increase in gain, due to lead, will increase ω_c slightly, and the phase margin becomes negative. To compensate, we need a large ω_a to make up for that lost phase and push up to $\sim 40^\circ$ phase margin. Thus, choose $\omega_a = 100$ placing zero at crossover, $\omega_c \approx 2$.

The plant phase rolls off quicker than the phase lead compensation gives you. So you want to push crossover as little as possible. Therefore, you want ω_c to occur early in the lead compensator:

This will reduce the bad effect of the system phase at higher frequencies.

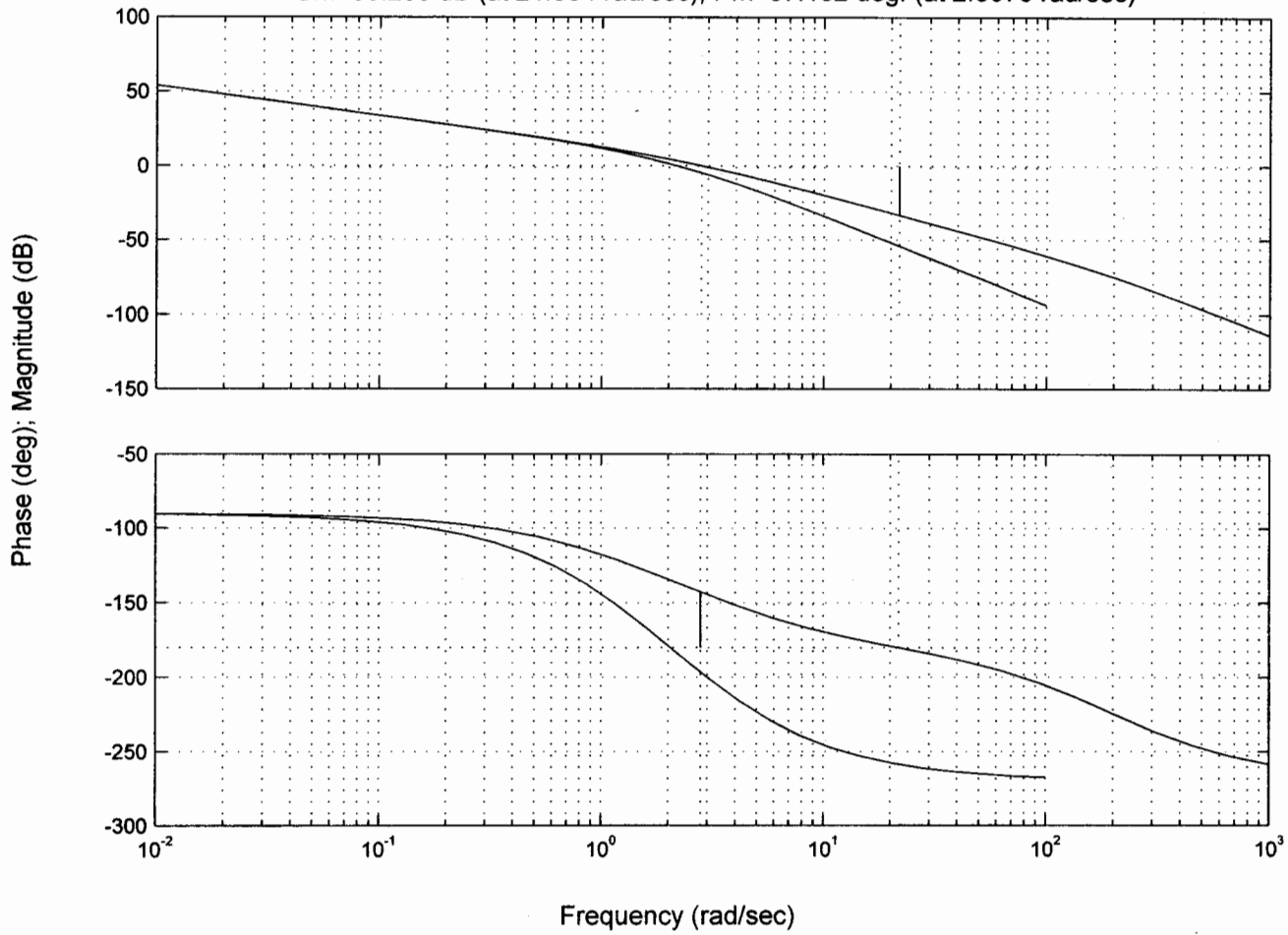


$$G_{\text{lead}} = \frac{s/2 + 1}{s/200 + 1} \quad \text{Try this compensator and iterate if necessary}$$

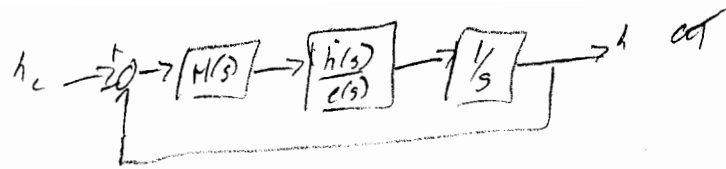
check bode plots ... Approx. bandwidth is open loop ω_c
 There are MANY answers! $\omega_c = 2.9076$

Bode Diagrams

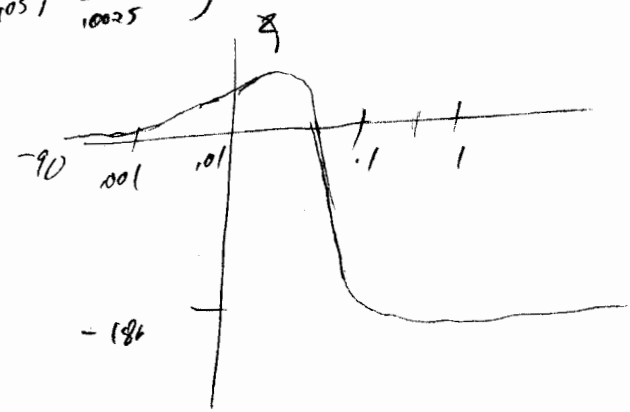
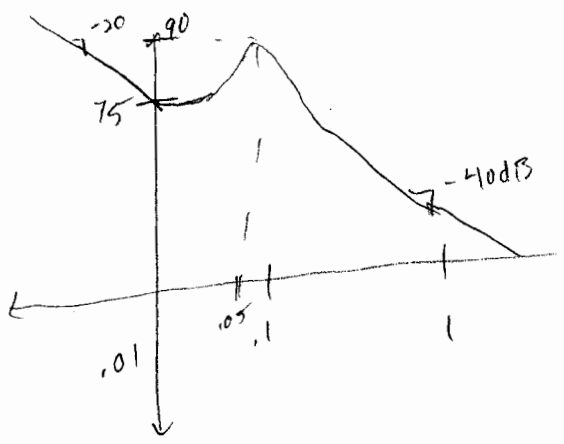
Gm=33.295 dB (at 21.884 rad/sec), Pm=37.132 deg. (at 2.8076 rad/sec)



2.) $\frac{h(s)}{e(s)} = \frac{15(s+0.01)}{s^2 + 0.01s + 0.0025}$
 a.) $\frac{h(s)}{e(s)} = \frac{15(s+0.01)}{s^2 + 0.01s + 0.0025}$



$\frac{h(s)}{e(s)} = \frac{15(s+0.01)}{s(s^2 + 0.01s + 0.0025)} = \frac{60 \left(\frac{s}{0.01} + 1\right)}{s \left(\left(\frac{s}{105}\right)^2 + \frac{0.01s}{10025} + 1\right)}$

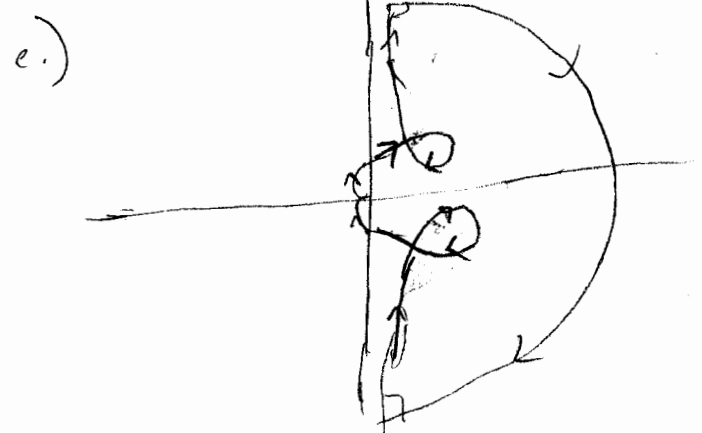


b.) $\left| \frac{h(s)}{e(s)} \right| = 1 = K \frac{60 \left(\left(\frac{1.16}{0.01} \right)^2 + 1 \right)^{1/2}}{0.16 \left(\left(\frac{0.01 \cdot 1.16}{0.0025} \right)^2 + \left(1 - \frac{1.16}{105} \right)^2 \right)^{1/2}}$

$K = \frac{1}{649} = 0.0015$

c.) Yes, the phase never goes below -180

d.) $\phi(s) = -90 - \tan^{-1} \left(\frac{1.16}{0.01} \right) - \tan^{-1} \left(\frac{0.01 \cdot 1.16}{0.0025 - 1.16^2} \right) = -179.615$
 $\phi_m = 0.385$



f.) $1 + K G(s) = (s^2 + 0.01s + 0.0025)s + K(15)(s+0.01) = 0$
 $K = 0.0015$
 Solve system for s

Poles at: $s = -0.009 \pm j0.158i$

$$9.) \frac{E(s)}{R(s)} = \frac{1}{1+6s} = \frac{1}{1+15(s+0.01)} = \frac{1}{s(s^2+0.01s+0.0025)} = \frac{s(s^2+0.01s+0.0025)}{s(s^2+0.01s+0.0025)+15(s+0.01)K}$$

$$c_{ss} = \lim_{s \rightarrow 0} (sE(s)) = \lim_{s \rightarrow 0} \left(s \cdot \frac{1}{s} \cdot \frac{s(s^2+0.01s+0.0025)}{s(s^2+0.01s+0.0025)+15(s+0.01)K} \right)$$

$$= \frac{0.0025}{15 \cdot 0.01 K} = \frac{1}{60 K} = \boxed{11.1 = c_{ss}}$$

w) $H(s) = K \frac{s/a + 1}{s/b + 1}$

PS

write $H(s) = K' \frac{\alpha Ts + 1}{Ts + 1}$ because know $\phi_H = \sin^{-1} \left(\frac{\alpha - 1}{\alpha + 1} \right)$ and $\omega_c = \frac{1}{\sqrt{\alpha} T}$

$\angle G(\omega = 0.16) = \tan^{-1} \left(\frac{0.16}{0.01} \right) - 90^\circ - \tan^{-1} \left(\frac{0.21}{0.005} \right) - \tan^{-1} \left(\frac{0.11}{0.005} \right) = -179.6^\circ$

choose $\alpha = 10$ so that $\phi_H \approx 55^\circ$

$T = \frac{1}{\sqrt{\alpha} \omega_c} = 1.98$

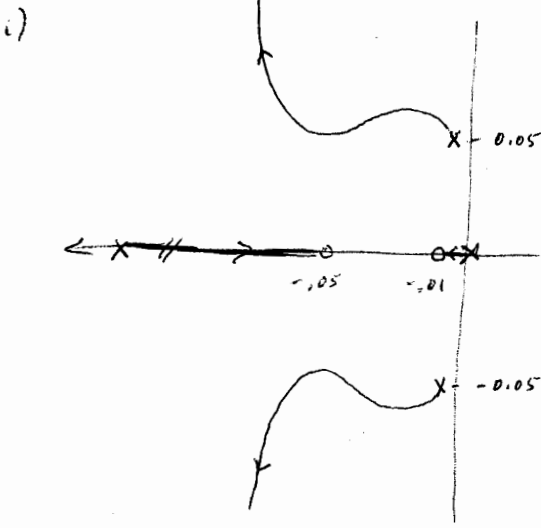
$a = \frac{1}{\alpha T} = 0.05$
 $b = \frac{1}{T} = 0.5$

need $K \left| \frac{(s/0.05 + 1)}{(s/0.5 + 1)} \right| |G(s)| = 1$ at $\omega_c = 0.16 \rightarrow K = 4.83 \times 10^{-4}$

~~other solutions are possible that would give a slightly different~~

Result is $PM = 55.3^\circ$ at $\omega_c = 0.16$

other solutions are possible



CL poles for K from part (h)

-0.0072
 -0.31
 $-0.097 \pm 0.083j$

j) for type 1 system

$$e_{ss} \text{ for a ramp input} = \frac{1}{K_{bode}} = \frac{1}{(60)(4.93 \times 10^{-4})} = 34.5$$

ps
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k) Need to increase the type of the system to type 2 to make error go to zero.

- Add integrator, $\frac{1}{s}$

However, the integrator will give too much negative phase.

- Add low frequency zero at least 1 decade before ω_c
 $\sim > \frac{1}{0.016}$

Compensator of form

$$\boxed{K \frac{(s+1)}{s}}$$